

SiC JFET Division

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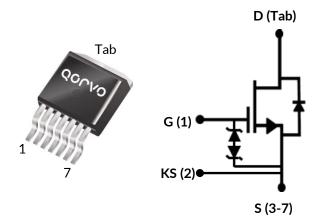








UF4C120070B7S



Part Number	Package	Marking
UF4C120070B7S	D ² PAK-7L	UF4C120070B7S







Silicon Carbide (SiC) Cascode JFET -EliteSiC, Power N-Channel, D2PAK-7L, 1200 V, 72 mohm

Rev B, January 2025

Description

The UF4C120070B7S is a 1200V, $72m\Omega$ G4 SiC FET. It is based on a unique 'cascode' circuit configuration, in which a normally-on SiC JFET is co-packaged with a Si MOSFET to produce a normally-off SiC FET device. The device's standard gate-drive characteristics allows use of off-the-shelf gate drivers hence requiring minimal re-design when replacing Si IGBTs, Si superjunction devices or SiC MOSFETs. Available in the space-saving D²PAK-7L package which enables automated assembly, this device exhibits ultra-low gate charge and exceptional reverse recovery characteristics, making it ideal for switching inductive loads and any application requiring standard gate drive.

Features

- On-resistance R_{DS(on)}: 72mΩ (typ)
- Operating temperature: 175°C (max)
- Excellent reverse recovery: Q_{rr} = 101nC
- ◆ Low body diode V_{FSD}: 1.43V
- Low gate charge: Q_G = 37.8nC
- Threshold voltage V_{G(th)}: 4.8V (typ) allowing 0 to 15V drive
- Low intrinsic capacitance
- ESD protected: HBM class 2 and CDM class C3
- D²PAK-7L package for faster switching, clean gate waveforms

Typical applications

- EV charging
- PV inverters
- Switch mode power supplies
- Power factor correction modules
- Induction heating













Maximum Ratings

Parameter	Symbol	Test Conditions	Value	Units
Drain-source voltage	V_{DS}		1200	V
Gate-source voltage	V	DC	-20 to +20	V
Gate-source voltage	$V_{GS} = \begin{array}{c} DC \\ AC (f > 1Hz) \\ T_{C} = 25^{\circ}C \\ \hline T_{C} = 100^{\circ}C \\ \hline I_{DM} \\ E_{AS} \\ dv/dt \\ \end{array}$	-25 to +25	V	
Continuous drain current ¹	I_	T _C = 25°C	25.7	Α
Continuous drain current	'D	T _C = 100°C	19.2	Α
Pulsed drain current ²	I _{DM}	T _C = 25°C	76	Α
Single pulsed avalanche energy ³	E _{AS}	L=15mH, I _{AS} =2.2A	36	mJ
SiC FET dv/dt ruggedness	dv/dt	V _{DS} ≤ 800V	200	V/ns
Power dissipation	P _{tot}	T _C = 25°C	183	W
Maximum junction temperature	$T_{J,max}$		175	°C
Operating and storage temperature	T_J, T_{STG}		-55 to 175	°C
Reflow soldering Temperature	T_{solder}	reflow MSL 1	245	°C

- 1. Limited by $T_{J,max}$
- 2. Pulse width t_p limited by $T_{J,max}$
- 3. Starting $T_J = 25$ °C

Thermal Characteristics

Parameter	Symbol Test Conditions		Value			Units
Parameter	Symbol	rest Conditions	Min	Тур	Max	Offits
Thermal resistance, junction-to-case	$R_{\theta JC}$			0.63	0.82	°C/W













Electrical Characteristics ($T_J = +25^{\circ}$ C unless otherwise specified) Typical Performance - Static

Parameter	Symbol Test Conditions		Value			Units
Parameter	Symbol	Test Conditions	Min	Тур	Max	Units
Drain-source breakdown voltage	BV _{DS}	V_{GS} =0V, I_D =1mA	1200			V
		V _{DS} =1200V,		0.4	10	
Total drain lookaga surrant		$V_{GS}=0V, T_J=25$ °C		0.4	18	
Total drain leakage current	I _{DSS}	V _{DS} =1200V,		40		- μΑ
		V _{GS} =0V, T _J =175°C		10		
Tatal cata leakage auguset		V _{DS} =0V, T _J =25°C,		,	20	
Total gate leakage current	I _{GSS}	V _{GS} =-20V / +20V		6		μΑ
		V _{GS} =12V, I _D =20A,		70	91	
		T _J =25°C		72		
Drain-source on-resistance	D	V _{GS} =12V, I _D =20A,		140		mΩ
Drain-source on-resistance	R _{DS(on)}	T _J =125°C		140		
		V_{GS} =12V, I_{D} =20A,		107		
		_{Т,} =175°С		197		
Gate threshold voltage	$V_{G(th)}$	V_{DS} =5V, I_{D} =10mA	4	4.8	6	V
Gate resistance	R_{G}	f=1MHz, open drain		4.5		Ω

Typical Performance - Reverse Diode

Parameter	Symbol Test Conditions –		Value			Units	
rarameter	Зуппрог	Test Conditions	Min	Тур	Max	Offics	
Diode continuous forward current ¹	I _S	T _C =25°C			25.7	Α	
Diode pulse current ²	$I_{S,pulse}$	T _C =25°C			76	Α	
	V	V_{GS} =0V, I_{S} =10A, T_{J} =25°C		1.43	1.64	.,	
Forward voltage	V _{FSD}	V_{GS} =0V, I_{S} =10A, T_{J} =175°C		2.38		V	
Reverse recovery charge	Q _{rr}	V_{DS} =800V, I_{S} =20A, V_{GS} =-5V, R_{G} =20 Ω ,		101		nC	
Reverse recovery time	t _{rr}	di/dt=1800A/μs, Τ _J =25°C		11		ns	
Reverse recovery charge	Q_{rr}	V_{DS} =800V, I_{S} =20A, V_{GS} =-5V, R_{G} =20 Ω ,		116		nC	
Reverse recovery time	t _{rr}	di/dt=1800A/μs, Τ _J =150°C		11		ns	













Typical Performance - Dynamic

D	6 1 1	T 16 12	Value			11.2
Parameter	Symbol	Test Conditions	Min	Тур	Max	- Units
Input capacitance	C _{iss}	- V _{DS} =800V, V _{GS} =0V -		1370		
Output capacitance	C_{oss}	f=100kHz		35		pF
Reverse transfer capacitance	C_{rss}	1-100KHZ		2		
Effective output capacitance, energy related	C _{oss(er)}	V_{DS} =0V to 800V, V_{GS} =0V		42		pF
Effective output capacitance, time related	C _{oss(tr)}	V _{DS} =0V to 800V, V _{GS} =0V		71		pF
C _{OSS} stored energy	E _{oss}	V _{DS} =800V, V _{GS} =0V		13.4		μЈ
Total gate charge	Q_{G}	V _{DS} =800V, I _D =20A,		37.8		
Gate-drain charge	Q_{GD}	$V_{DS}=000V$, $I_D=20\Delta$, $V_{GS}=0V$ to 15V		9.5		nC
Gate-source charge	Q_GS	V GS-0V 10 13 V		10		
Turn-on delay time	$t_{d(on)}$			20		
Rise time	t _r	Notes 4 and 5, $V_{DS}=800V, I_{D}=20A, Gate$ $Driver=-5V \text{ to } +15V,$ $R_{G_{ON}}=10\Omega, R_{G_{OFF}}=50\Omega,$ inductive Load, $FWD: \text{same device with}$		32		ns
Turn-off delay time	t _{d(off)}			57		
Fall time	t _f			12		
Turn-on energy including R _S energy	E _{ON}			352		μ
Turn-off energy including R _S energy	E _{OFF}	V_{GS} =-5V and R_{G_OFF} =50 Ω ,		62		
Total switching energy	E _{TOTAL}	Device Snubber:		414		
Snubber R _S energy during turn-on	E _{RS_ON}	Cs=100pF,Rs=20Ω T ₁ =25°C		10		
Snubber R _S energy during turn-off	E _{RS_OFF}]		8		
Turn-on delay time	$t_{d(on)}$	Notes 4 and 5,		24		
Rise time	t _r	V _{DS} =800V, I _D =20A, Gate		33		nc
Turn-off delay time	$t_{d(off)}$	Driver=-5V to +15V,		63		– ns
Fall time	t_f	$R_{G_ON}=10\Omega$, $R_{G_OFF}=50\Omega$, inductive Load,		13		
Turn-on energy including R _S energy	E _{ON}	FWD: same device with		396		
Turn-off energy including R _S energy	E _{OFF}	V_{GS} =-5V and R_{G_OFF} =50 Ω ,		81		
Total switching energy	E _{TOTAL}	Device Snubber:		477		μJ
Snubber R _S energy during turn-on	E _{RS_ON}	Cs=100pF,Rs=20Ω T ₁ =150°C		5		
Snubber R _S energy during turn-off	E_{RS_OFF}	1 _J =130 C		12		

^{4.} Measured with the half-bridge mode switching test circuit in Figure 26.

^{5.} In this datasheet, all the switching energies (turn-on energy, turn-off energy and total energy) presented in the tables and Figures include the device RC snubber energy losses.













Typical Performance Diagrams

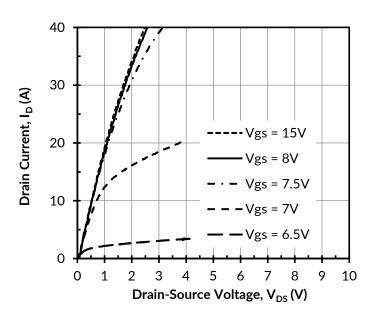


Figure 1. Typical output characteristics at T_J = - 55°C, tp < 250 μ s

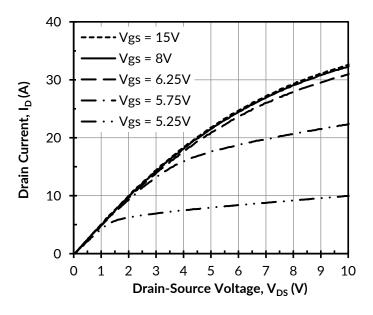


Figure 3. Typical output characteristics at T_J = 175°C, tp < 250 μ s

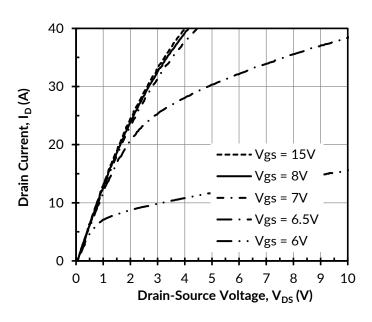


Figure 2. Typical output characteristics at T_J = 25°C, tp < 250 μ s

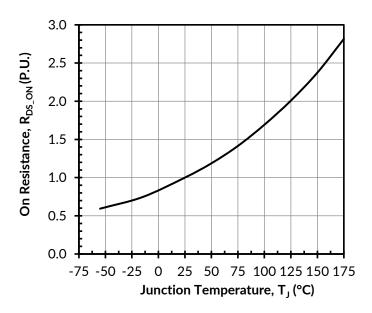


Figure 4. Normalized on-resistance vs. temperature at V_{GS} = 12V and I_D = 20A



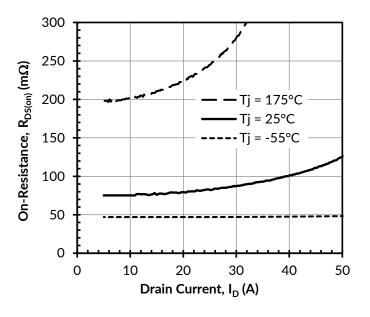








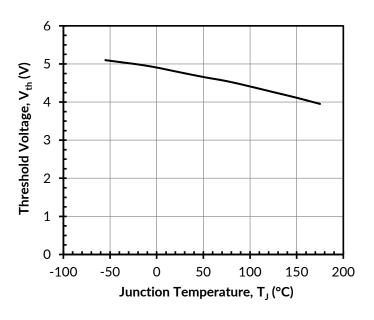




30 Tj = -55°C 25 Tj = 25°C Tj = 175°C Drain Current, I_D (A) 20 15 10 5 0 5 0 1 3 6 9 10 2 Gate-Source Voltage, V_{GS} (V)

Figure 5. Typical drain-source on-resistances at V_{GS} = 12V

Figure 6. Typical transfer characteristics at V_{DS} = 5V



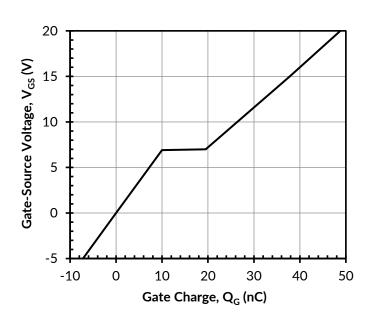


Figure 7. Threshold voltage vs. junction temperature at V_{DS} = 5V and I_{D} = 10mA

Figure 8. Typical gate charge at I_D = 20A and V_{DS} =800V



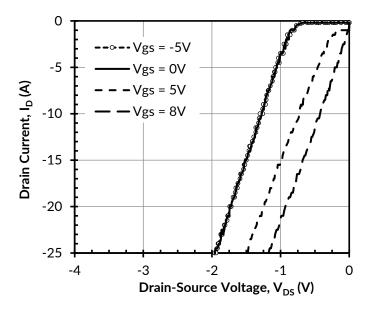








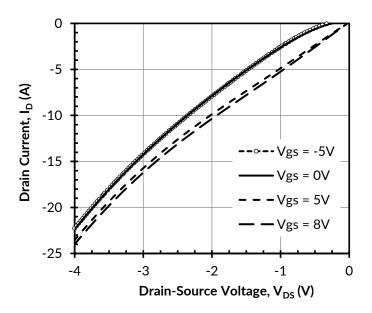




0 --- Vgs = - 5V Vgs = 0V -5 - Vgs = 5V Drain Current, I_D (A) **-** Vgs = 8V -10 -15 -20 -25 -3 -1 -2 0 Drain-Source Voltage, V_{DS} (V)

Figure 9. 3rd quadrant characteristics at $T_J = -55$ °C

Figure 10. 3rd quadrant characteristics at $T_J = 25$ °C



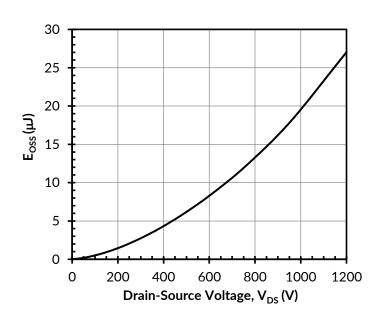


Figure 11. 3rd quadrant characteristics at $T_J = 175$ °C

Figure 12. Typical stored energy in C_{OSS} at $V_{GS} = 0V$



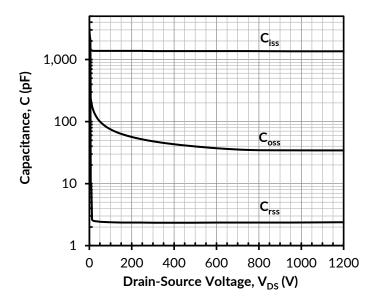












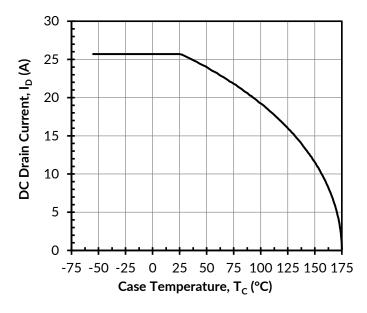


Figure 13. Typical capacitances at f = 100kHz and V_{GS} = 0V

Figure 14. DC drain current derating

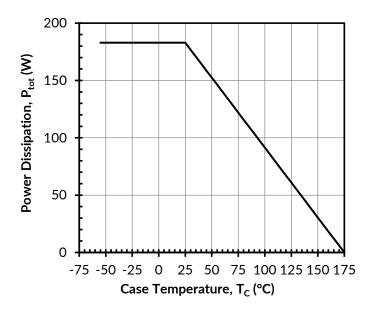


Figure 15. Total power dissipation

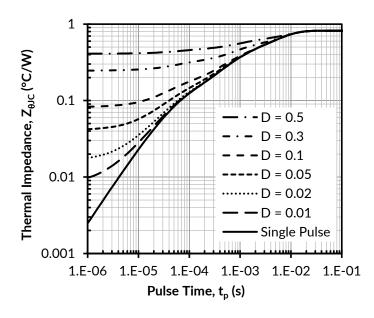


Figure 16. Maximum transient thermal impedance













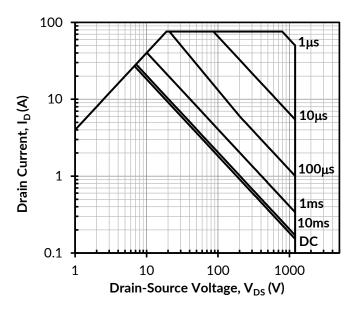


Figure 17. Safe operation area at T_C = 25°C, D = 0, Parameter t_p

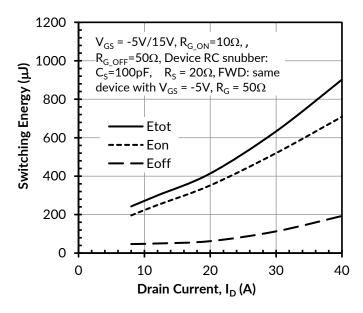


Figure 19. Clamped inductive switching energy vs. Drain Current at V_{DS} = 800V and T_J = 25°C

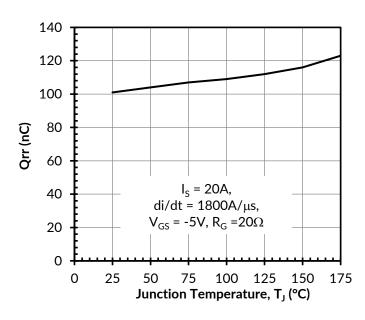


Figure 18. Reverse recovery charge Qrr vs. junction temperature at V_{DS} = 800V

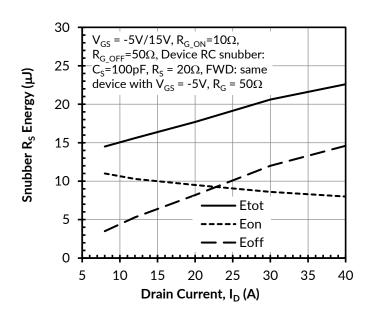


Figure 20. RC snubber energy loss vs. Drain Current at $V_{DS} = 800V$, $I_D = 20A$, and $T_J = 25$ °C













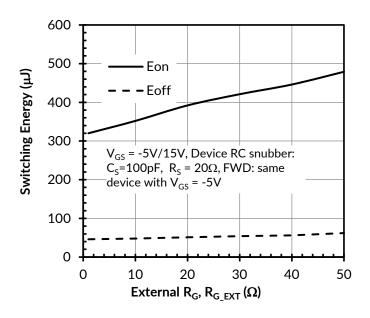


Figure 21. Clamped inductive switching energies vs. $R_{G,EXT}$ at V_{DS} = 800V, I_{D} = 20A, and T_{J} = 25°C

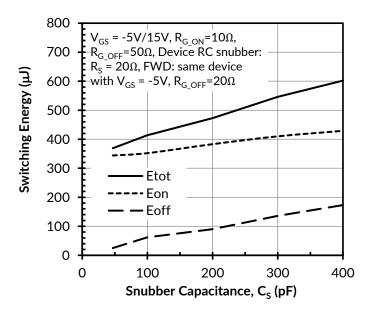


Figure 23. Clamped inductive switching energies vs. snubber capacitance C_S at V_{DS} = 800V, I_D = 20A, and T_1 = 25°C

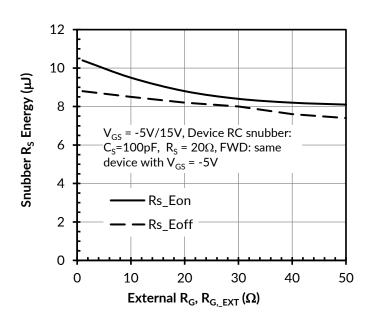


Figure 22. RC snubber energy loss vs. $R_{G,EXT}$ at V_{DS} = 800V, I_D = 20A, and T_J = 25°C

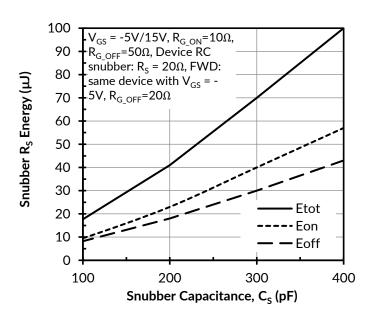


Figure 24. RC snubber energy losses vs. snubber capacitance C_S at V_{DS} = 800V, I_D = 20A, and T_J = 25°C



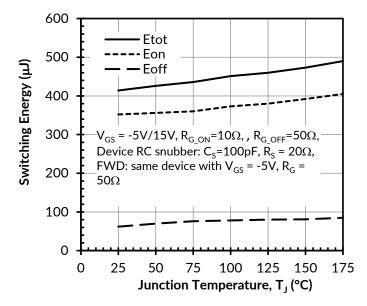












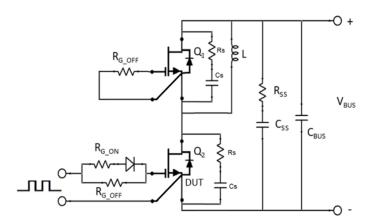


Figure 25. Clamped inductive switching energy vs. junction temperature at V_{DS} = 800V and I_D =20A

Figure 26. Schematic of the half-bridge mode switching test circuit with device RC snubbers (R_s = 20 Ω , C_s = 100pF) and a bus RC snubber (R_{SS} = 2.5 Ω , C_{SS} =100nF).

Applications Information

SiC FETs are enhancement-mode power switches formed by a high-voltage SiC depletion-mode JFET and a low-voltage silicon MOSFET connected in series. The silicon MOSFET serves as the control unit while the SiC JFET provides high voltage blocking in the off state. This combination of devices in a single package provides compatibility with standard gate drivers and offers superior performance in terms of low on-resistance ($R_{DS(on)}$), output capacitance (C_{oss}), gate charge (Q_G), and reverse recovery charge (Q_G) leading to low conduction and switching losses. The SiC FETs also provide excellent reverse conduction capability eliminating the need for an external anti-parallel diode.

Like other high performance power switches, proper PCB layout design to minimize circuit parasitics is strongly recommended due to the high dv/dt and di/dt rates. An external gate resistor is recommended when the FET is working in the diode mode in order to achieve the optimum reverse recovery performance. For more information on SiC FET operation, see https://www.qorvo.com/design-hub.

A snubber circuit with a small $R_{(G)}$, or gate resistor, provides better EMI suppression with higher efficiency compared to using a high $R_{(G)}$ value. There is no extra gate delay time when using the snubber circuitry, and a small $R_{(G)}$ will better control both the turn-off $V_{(DS)}$ peak spike and ringing duration, while a high $R_{(G)}$ will damp the peak spike but result in a longer delay time. In addition, the total switching loss when using a snubber circuit is less than using high $R_{(G)}$, while greatly reducing $E_{(OFF)}$ from mid-to-full load range with only a small increase in $E_{(ON)}$. Efficiency will therefore improve with higher load current. For more information on how a snubber circuit will improve overall system performance, visit the Qorvo website at https://www.qorvo.com/design-hub.













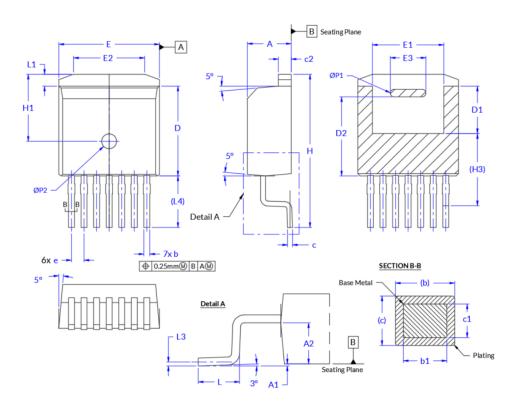
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TO263-7L (D2PAK-7L) PACKAGE OUTLINE, MARKING, TAPE AND REEL SPECIFICATION	PART	Page 1 of 4
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PACKAGE OUTLINE



	7L-D2PAK				
SYM	М	М	IN	CH	
31141	Min	Max	Min	Max	
Α	4.30	4.56	.169	.180	
A1	0.00	0.25	.000	.010	
A2	2.45	2.75	.096	.108	
b	0.50	0.70	.020	.028	
b1	0.50	-	.020	-	
С	0.40	0.60	.016	.024	
c1	0.40		.016		
c2	1.20	1.40	.047	.055	
D	8.93	9.23	.352	.363	
D1	4.65	4.95	.183	.195	
D2	7.90	8.10	.311	.319	
e	1.27	BSC	.050 BSC		
E	10.08	10.28	.397	.405	
E1	6.82	7.62	.269	.300	
E2	6.50	8.60	.256	.339	
E3	3.50	3.70	.138	.146	
Н	15.00	16.00	.591	.630	
H1	6.68	6.88	.263	.271	
H3	7.31	REF.	.287	REF	
L	1.90	2.50	.075	.098	
L1	0.98	1.42	.039	.056	
L3	0.25	BSC	.0098	BSC	
L4	5.22	REF	.205	REF	
ØP1	0.65	0.85	.026	.033	
ØP2	1.40	1.60	.055	.063	

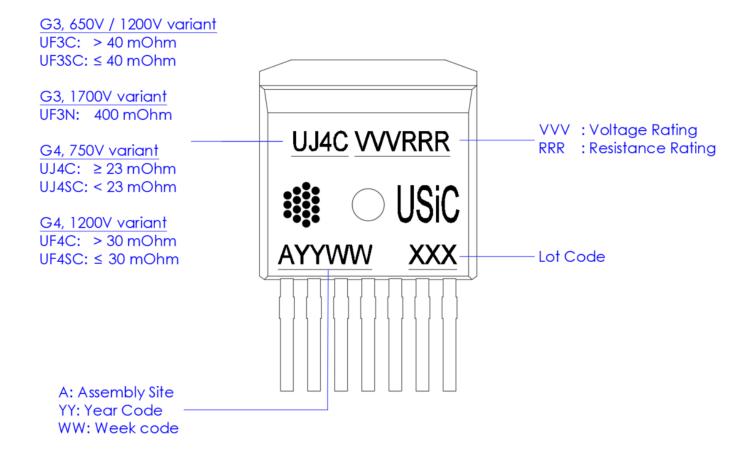
Notes:

- 1. GENERAL TOLERANCE: ±0.1 unless otherwise specified
- 2. CONTROLLING DIMENSION: MILLIMETERS
- 3. PACKAGE BODY SIDES EXCLUDE MOLD FLASH AND GATE BURRS.
- 4. DIMENSION LIS MEASURED IN GAUGE LINE.
- 5. CONVERTED INCH DIMENSIONS ARE NOT NECESSARILY EXACT.
- 6. DIMENSION c1 AND b1 APPLIES TO BASE METAL ONLY



TO263-7L (D2PAK-7L) PACKAGE OUTLINE, PART MARKING, TAPE AND REEL SPECIFICATION	Page 2 of 4
DS_TO_263_7L	Rev D

PART MARKING



Template: FOR-000530 Rev G



TO263-7L	(D2PAK-7L)	PACKAGE	OUTLINE,	PART
MARKING,	TAPE AND RE	EL SPECIFIC	ATION	

TO 000 7

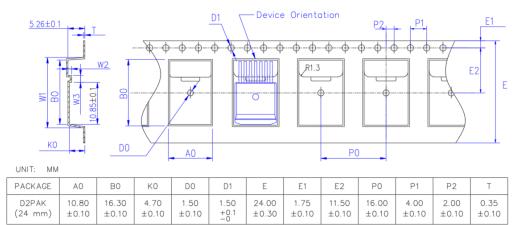
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Rev D

DS_TO_263_7L

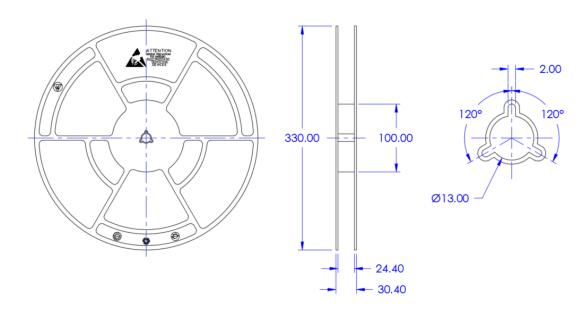
PACKING TYPE

Carrier Tape



Ext	erior	size	
	W1	16.9±0.1	
Spec	W2	1.3±0.1	
'	W3	1.0±0.1	
	W1	17.2±0.1	(1)
Spec 2	W2	1.8±0.1	(b)
	W3	0.85±0.1	0

Reel



All dimensions in millimeters Anti-Static Tape and Rell (T&R) Quantity per Reel: 800 units



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REVISION HISTORY

Revision	Create Date (mm/dd/yyyy)	Description of Change	Initiator of Change
С	11/06/2023	Updated to Qorvo template Updated Package outline drawing based latest drawing revision	Glenn Galang
D	05/21/2024	Added illustration of device orientation on carrier tape (page 3)	Glenn Galang

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